

Simple conditions for L_2 stability and stabilization of networked control systems

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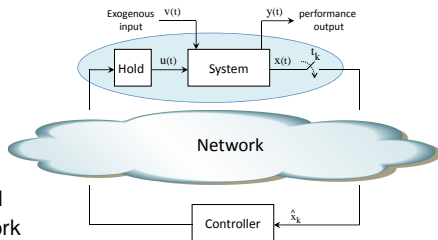
KTH, Stockholm, Sweden

IFAC World Congress 2011, Milano, Italy



Outline

- ▶ Introduction
- ▶ Problem statement
- ▶ Stability analysis
- ▶ Control
- ▶ Conclusion and Future Works



- ▶ Remote control
- ▶ Wireless network
 - ▶ Data loss
 - ▶ Time-varying propagation delays
 - ▶ Varying sampling period



Existing approaches

- ▶ Time-delay systems [Yu et al.], [Fridman et al.]
- ▶ Impulsive systems [Naghshtabrizi et al.], [Seuret]
- ▶ Sampled-data systems [Mirkin]
- ▶ Robust techniques [Fujioka], [Oishi et al.]
- ▶ Functional-based approaches [Seuret]



Problem statement



NCS model

► Process model

$$\begin{aligned} \dot{x}(t) &= Ax(t) + Bu(t) + Ev(t) \\ y(t) &= Cx(t) + Du(t) + Fv(t) \\ x(0) &= x_0 \end{aligned} \quad (1)$$

► Control-law model

$$\begin{aligned} u(t) &= Kx(t_k) \\ t &\in [t_k, t_{k+1}) \\ t_{k+1} - t_k &\leq (1+m)T_{max} + \tau_{k+1} \\ \tau_k &\in [0, \tau_{max}] \end{aligned} \quad (2)$$

- t_k : arrival instants of a new control input
 - $\tau_k = \tau(t_k)$, $k \in \mathbb{N}$
 - m is the number of consecutive dropouts
- Varying sampling period \rightarrow Actual (varying) sampling period+data loss+varying propagation delays



Problem

- ▶ Find a sampled-data state-feedback control law such that the closed-loop system is asymptotically (exponentially) stable and
 - objective 1:** maximize the Maximal Allowable Transfer Interval (MATI) under L_2 disturbance attenuation constraints; or
 - objective 2:** minimize L_2 disturbance attenuation gain under a MATI constraint.

$$\mathcal{S}_{\text{MATI}} := \{(m, \tau, T) \in \mathbb{N} \times \mathbb{R}_+ \times \mathbb{R}_{++} : (1 + m)T + \tau \leq \text{MATI}\}. \quad (3)$$

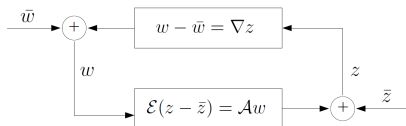
- ▶ m : number of consecutive dropouts
- ▶ T : sampling period
- ▶ τ : propagation delay



Stability Analysis



Quadratic separation



Theorem

The interconnected system above is well-posed if there exists a symmetric matrix Θ satisfying the conditions

$$\begin{bmatrix} \mathcal{E} & -\mathcal{A} \end{bmatrix}_{\perp}^T \Theta \begin{bmatrix} \mathcal{E} & -\mathcal{A} \end{bmatrix}_{\perp} \succ 0 \quad (4)$$

and

$$\left\langle \begin{bmatrix} 1 \\ \mathbb{P}_T \nabla \end{bmatrix} u_T, \Theta \begin{bmatrix} 1 \\ \mathbb{P}_T \nabla \end{bmatrix} u_T \right\rangle \leq 0 \quad (5)$$

for all $u \in L_{2e}$ and all $T > 0$.



Alternative system representation

$$\underbrace{\begin{bmatrix} x(t) \\ \delta(t) \\ v(t) \end{bmatrix}}_{w(t)} = \underbrace{\begin{bmatrix} \mathcal{I}1_n & & \\ & \Delta_{sh}1_n & \\ & & \Delta_\gamma \end{bmatrix}}_{\nabla} \underbrace{\begin{bmatrix} \dot{x}(t) \\ \dot{x}(t) \\ y(t) \end{bmatrix}}_{z(t)}, \quad (6)$$

$$\underbrace{\begin{bmatrix} 1 & 0 & 0 \\ -1 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}}_{\mathcal{E}} \underbrace{\begin{bmatrix} \dot{x}(t) \\ \dot{x}(t) \\ y(t) \end{bmatrix}}_{z(t)} = \underbrace{\begin{bmatrix} A+BK & -BK & E \\ 0 & 0 & 0 \\ C+DK & -DK & F \end{bmatrix}}_{\mathcal{A}} \underbrace{\begin{bmatrix} x(t) \\ \delta(t) \\ v(t) \end{bmatrix}}_{w(t)} \quad (7)$$

- ▶ \mathcal{I} : integral operator,
- ▶ $\Delta_{sh} : \theta \rightarrow \int_{t_k}^t \theta(s)ds, t \leq t_{k+1} \Rightarrow \delta(t) = x(t) - x(t_k).$
- ▶ Δ_γ : virtual operator characterizing the L_2 gain of the transfer $v \rightarrow y$.



IQC for the integral operator

Lemma

The integration operator \mathcal{I} is characterized by the IQC:

$$\left\langle \begin{bmatrix} \mathbf{1}_n \\ \mathcal{I}\mathbf{1}_n \end{bmatrix} x_T, \begin{bmatrix} 0 & -P \\ -P & 0 \end{bmatrix} \begin{bmatrix} \mathbf{1}_n \\ \mathcal{I}\mathbf{1}_n \end{bmatrix} x_T \right\rangle \leq 0.$$

for all $x \in L_{2e}^n$ and for any matrix $P \in \mathbb{S}_{++}^n$.

- ▶ Lyapunov condition for stability
- ▶ Frequency domain $\mathcal{I} \rightarrow s^{-1}$

$$\begin{bmatrix} \mathbf{1}_n \\ s^{-1}\mathbf{1}_n \end{bmatrix}^* \begin{bmatrix} 0 & -P \\ -P & 0 \end{bmatrix} \begin{bmatrix} \mathbf{1}_n \\ s^{-1}\mathbf{1}_n \end{bmatrix} \preceq 0, \forall \Re[s] \geq 0$$

Pre- and post-multiply by $s^*X(s)^*$ and $sX(s)$

$$-(s + s^*)X(s)^*PX(s) \preceq 0, \forall \Re[s] \geq 0 \quad (8)$$

Characterization of all $\Re[s] \geq 0$: pick any $P = P^T \succ 0$





IQC for Δ_{sh}

Lemma

The operator Δ_{sh} can be characterized by the IQC:

$$\left\langle \begin{bmatrix} \mathbf{1}_n \\ \Delta_{sh} \mathbf{1}_n \end{bmatrix} x_T, \begin{bmatrix} -\frac{4}{\pi^2} \mu^2 S_1 & -S_2 \\ -S_2 & S_1 \end{bmatrix} \begin{bmatrix} \mathbf{1}_n \\ \Delta_{sh} \mathbf{1}_n \end{bmatrix} x_T \right\rangle \leq 0.$$

for all $x \in L_{2e}^n$ and for any matrices $S_1, S_2 \in \mathbb{S}_{++}^n$.

- ▶ S_1 : Bound on the L_2 -gain of $2\mu/\pi$, $t_{k+1} - t_k \leq \mu$, $k \in \mathbb{N}$ [Mirkin]
- ▶ S_2 : Passivity of Δ_{sh} [Fujioka]



IQC for Δ_γ

Lemma

The operator Δ_γ , which has an L_2 -induced norm equal to γ^{-1} , is characterized by the IQC:

$$\left\langle \begin{bmatrix} \mathbf{1}_r \\ \Delta_\gamma \end{bmatrix} x_T, \begin{bmatrix} -\gamma^{-2} \mathbf{1}_q & 0 \\ 0 & \mathbf{1}_r \end{bmatrix} \begin{bmatrix} \mathbf{1}_q \\ \Delta_\gamma \end{bmatrix} x_T \right\rangle \leq 0,$$

for all $x \in L_{2e}^q$.

- ▶ Upper bound γ on L_2 -gain of $v \rightarrow y$



Stability result

Theorem

The NCS system is asymptotically stable for all $(m, \tau, T) \in \mathcal{S}_\mu$ if there exist matrices $P, S_1, S_2 \in \mathbb{S}_{++}^n$ and a scalar $\eta > 0$ such that the LMI

$$\begin{bmatrix} \mathcal{E} & -\mathcal{A} \end{bmatrix}_\perp^T \Theta \begin{bmatrix} \mathcal{E} & -\mathcal{A} \end{bmatrix}_\perp \prec 0 \quad (9)$$

holds where \mathcal{E}, \mathcal{A} are defined in (7) and

$$\Theta = \left[\begin{array}{ccc|ccc} 0 & 0 & 0 & -P & 0 & 0 \\ 0 & -\frac{4}{\pi^2} \mu^2 S_1 & 0 & 0 & -S_2 & 0 \\ 0 & 0 & -\eta \mathbf{1}_q & 0 & 0 & 0 \\ \hline & * & & 0 & 0 & 0 \\ & & & 0 & S_1 & 0 \\ & & & 0 & 0 & \mathbf{1}_r \end{array} \right]. \quad (10)$$

Moreover, the closed-loop system satisfies $\|y\|_{L_2} \leq \sqrt{1/\eta} \|v\|_{L_2}$.



Example

$$\dot{x}(t) = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} [-1.006 \quad -1.006] x(t_k). \quad (11)$$

- ▶ Maximal constant sampling period: 5.8117.

	<i>MATI</i>	nb. of vars.	for $n = 2$
[Yu, 04]	unfeasible	$4 \frac{n(n+1)}{2}$	12
[Yue, 04]	0.970	$2 \frac{n(n+1)}{2} + 6n^2$	30
[Tan, 08]	0.995	$4 \frac{n(n+1)}{2} + 16n^2$	76
[Naghshtabrizi, 06](without delay)	1.272	$7 \frac{n(n+1)}{2} + 16n^2$	85
Proposed result	1.561	$3 \frac{n(n+1)}{2} + 1$	10



Stabilization



Main result

Theorem

There exists a matrix $K \in \mathbb{R}^{m \times n}$ such that the NCS is asymptotically stable for all $(m, \tau, T) \in \mathcal{S}_\mu$ if there exist matrices $P, S_1 \in \mathbb{S}_{++}^n$, $X \in \mathbb{R}^{n \times n}$, $U \in \mathbb{R}^{m \times n}$ and a scalar $\gamma > 0$ such that the LMI

$$\begin{bmatrix} -(X + X^T) & P + A'_{cl} & -BU & E & 0 & X & \mu \frac{\pi}{2} S_1 \\ \star & -P & 0 & 0 & C'_{cl}{}^T & 0 & 0 \\ \star & \star & -S_1 & 0 & -(DU)^T & 0 & 0 \\ \star & \star & \star & -\gamma I & F^T & 0 & 0 \\ \star & \star & \star & \star & -\gamma I & 0 & 0 \\ \star & \star & \star & \star & \star & -P & -\mu \frac{\pi}{2} S_1 \\ \star & \star & \star & \star & \star & \star & -S_1 \end{bmatrix} \prec 0 \quad (12)$$

holds with $A'_{cl} = AX + BU$ and $C'_{cl} = CX + DU$. Furthermore, the closed-loop system controlled with gain $K = UX^{-1}$ satisfies $\|y\|_{L_2} \leq \gamma \|v\|_{L_2}$.



Example

- ▶ Let us consider the open-loop system

$$\dot{x}(t) = \begin{bmatrix} -0.8 & -0.01 \\ 1 & 0.1 \end{bmatrix} x(t) + \begin{bmatrix} 0.4 \\ 0.1 \end{bmatrix} u(t) \quad (13)$$

- ▶ [Yu, 04]: system stabilizable for $\mu \leq 0.6011$.
- ▶ Proposed result: system stabilizable for $\mu \leq 3.64826$ with the controller gain $K = \begin{bmatrix} -0.3482 & -0.3097 \end{bmatrix}$.



Conclusion

- ▶ Approach based on well-posedness and IQCs
- ▶ LMI results for both stability and stabilization
- ▶ Tradeoff: L_2 -gain minimization vs. MATI maximization
- ▶ Low numerical complexity
- ▶ Can be extended to robust stability and stabilization



Thank you for your attention